# The Promise of Passively Cooled 10 µm HgCdTe Arrays for NGST

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# Requirements imposed by Passively Cooled Focal Planes

#### Why passively cooled focal planes?

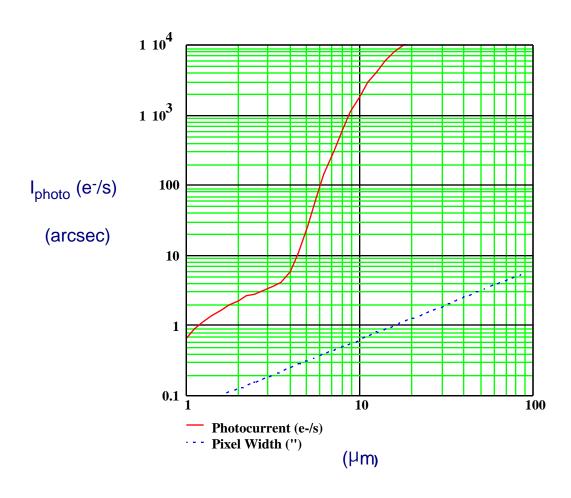
- No mechanical cooler required (lower weight, power & complexity, hence expense)
- Extended missions enabled
- Temperatures between 25 and 30K attainable
- "Faster, Better, Cheaper" space missions

#### Requirements on 10 µm HgCdTe Arrays

- Goal: < zodiacal background current (shown for assumed 4-m telescope, diffraction limited pixels, QE = 70%, telescope and camera efficiencies 81% and 58% respectively)
- Competitive dark currents possible if g-r limited

July 10, 1997

# South Ecliptic Pole Zodiacal Background Photocurrent (function of )



### Requirements on HgCdTe Detectors

#### Fundamental Dark Current Limitations

- For operating temperatures 30K, we model g-r limited dark currents for lifetimes of  $10^{-6}$  s to  $10^{-7}$  s at  $10.4~\mu m$  of 60 to 600 e<sup>-</sup>/s for doping density  $N_D = 10^{15}$  cm<sup>-3</sup> and 110 to 1100 e<sup>-</sup>/s for  $N_D = 10^{14}$  cm<sup>-3</sup>
- g-r limited  $R_o$ A's range from  $10^{8 \text{ to } 9}$  ohm-cm<sup>2</sup> at 30K to 2  $10^{10 \text{ to } 11}$  ohm-cm<sup>2</sup> at 25K!
- in practice, tunneling limits dark current

### Challenge

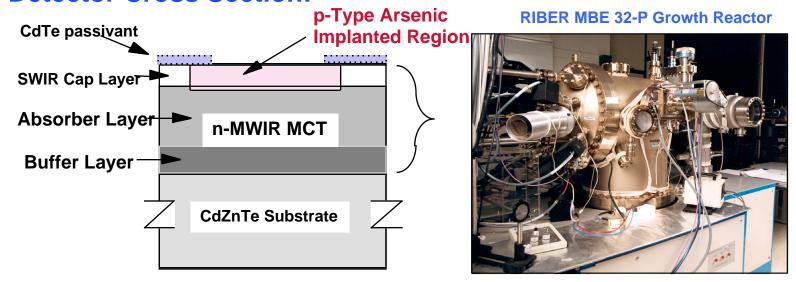
- Produce 10  $\mu$ m HgCdTe detectors which exhibit improved dark current at ~30K i.e. reduce  $N_D$  and reduce tunneling contributions to dark current
- At the same time, the QE must remain sufficiently high at this low temperature

### Rockwell HgCdTe DLPH Photovotaic Detectors

- $Hg_{1-x}Cd_xTe$ 
  - tunable band gap with composition x
  - high QE and instrinsically rad-hard
  - small lattice mismatch between HgCdTe and CdZnTe allows relatively high quality epitaxial heterostructures
- Double Layer Planar Heterostructure (DLPH)
  - in DLPH, the LWIR material is buried beneath a MWIR material and is never exposed to processing chemicals
  - immediately deposited MWIR effectively passivates the surface of the LWIR layer, eliminating the leakage where p/n junction intersects the LWIR/MWIR interface
  - Originally, Rockwell used an LPE process: they have now developed a superior MBE DLPH process which is sufficiently mature that there is good material control, low defect density, and good surface characteristics

## P/N Double Layer Planar Heterostructure-Offers Performance and Manufacturing Advantages

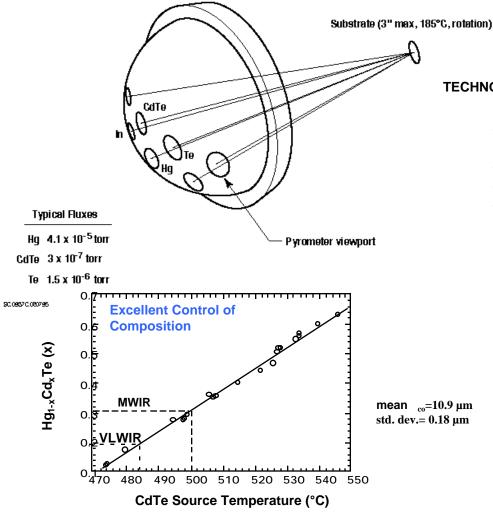
#### **Detector Cross Section:**



- Inherently High Radiation Hardness
- Inherently Easy to Passivate
- Few Critical Steps Give Reliable Processes
- Excellent Device Performance
- Exploits MBE's Outstanding Materials Control

#### Molecular Beam Epitaxy HgCdTe Growth Technology-

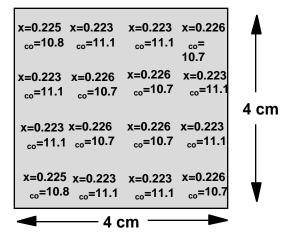
### **How it is Done**



#### **TECHNOLOGY ATTRIBUTES:**

Process Control Refined Due to
In-Situ Monitoring
Single System Services all IR Applications
Large Area Uniformity
Compositional Control Allows Advanced
Structures

#### **LWIR Material Composition Uniformity**



Test & Qualification

**FPA** 

# FPA DETECTOR DESIGN AND FABRICATION PERFORMED IN ROCKWELL HgCdTe LINE

#### **FABRICATION ROADMAP** MBE dZnTe Layer Array **Detector Double Layer** Screen **Substrate Processing** Mask Growth Wafer **Detector Die Detector - MCT Line Screen Test Selection Packaging ROIC Die** Design/ **ROIC** with BCS **Selection** Layout Fab

**ROIC** 

**Screen Test** 

**ROIC** 

**Processing** 

Readout

### Test Diode Philosophy

#### Extrapolate from other Rockwell Programs

- Rockwell developing DLPH arrays for low background applications with good 40K performance at > 10 µm
- extrapolate from those programs for our lower background applications

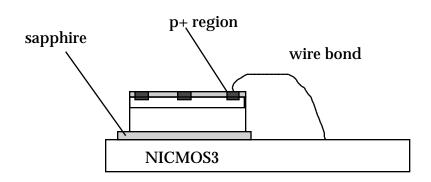
#### • Use of NICMOS3 Multiplexer

- Rochester test system set up for low-noise array measurements
- Bonded diodes (e.g. delivery #4, 18-diode wafer) to NICMOS mux with 1 mil wire bonds to Indium bumps on mux unit cell inputs
- Typically one diode connected to 3 5 mux pixels

#### **Detectors Received from Rockwell**

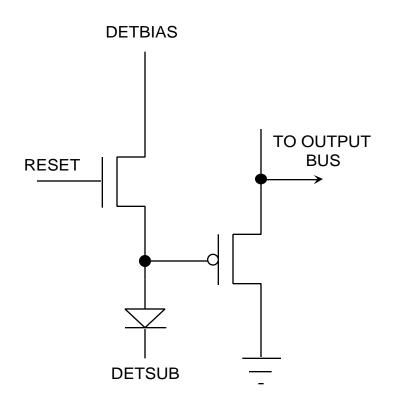
- Delivery 2: 13.7  $\mu$ m LPE, 5 measurable diodes,  $N_D = 1 \times 10^{15}$  cm<sup>-3</sup>
  - test results are consistent with Rockwell's results
  - dark current at 50 mV back-bias ~ $10^7$ e<sup>-</sup>/sec, best R<sub>0</sub>A~ $2 \times 10^7$  -cm<sup>2</sup>
- Delivery 3: 10.3  $\mu$ m MBE, 8 measurable diodes,  $N_D = 1.7 \times 10^{15} \text{ cm}^{-3}$ 
  - because of the high doping density, the dark current performance is not as good as Delivery 2. Also shows large diode to diode variance.
    - » dark current has large bias dependence, small temperature dependence - a signature of tunneling
- Delivery 4: 10.6  $\mu$ m MBE, 17 measurable diodes;  $N_D = 9.5 \times 10^{14} \text{ cm}^{-3}$ . This delivery will be discussed here.
- Delivery 5: 10  $\mu$ m MBE,  $N_D = 6.4 \times 10^{14}$  cm<sup>-3</sup>; 256 x 256 array. Indium-bump-bonded to TCM2620 multiplexer tests in progress

### **Test Structure for Deliveries 1-4**



- Diodes are wire-bonded to NICMOS3 muxes
  - advantages
    - low noise since long integration times and multiple sampling: good for small signals
    - » fast I-V curve test
  - disadvantage
    - » front illumination
    - » integration capacitance hard to estimate (multinodes/detector)

# Finding the NICMOS Mux Nodes Which Are Connected to the Diodes



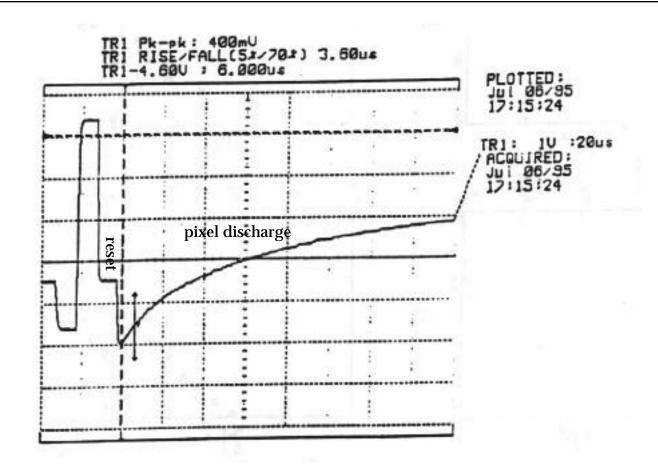
 $V_{bias} = V_{DETSUB} - V_{DETBIAS}$ 

- Set  $V_{DETBIAS}$ =-500 mV
- Difference of images in uncorrelated single sampling mode with  $V_{bias}$ =300 mV and  $V_{bias}$ =-200 mV were taken
- The pixels with response (> 2500 ADUs) are the pixels bonded to the diodes
- Made sure all the responding pixels discharge when the reset switches were turned off.

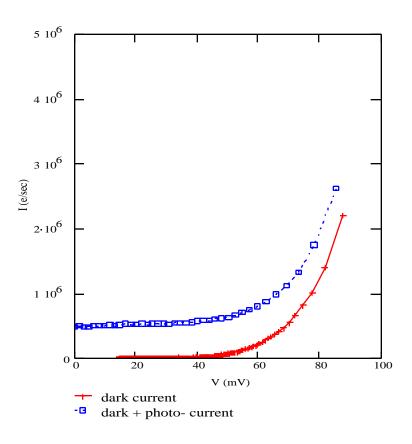
#### **I-V Curve Test**

- Set  $V_{bias}$ = 100 mV (actual back bias is about 170 mV, *i.e.* the zero bias point is -70 mV)
- Select pixel to be measured and reset it
- Turn reset switch off and record the discharging of the output level as a function of time (with 16-bit A/D converter). Oscilloscope trace of discharge next slide.
- To deduce the I-V curve
  - V=(output level-totally discharged level)/(mux gain x preamp gain)
  - $I=(C \times [d(output level)/dt])/(mux gain \times preamp gain)$ 
    - » capacitance C is the integration capacitance, deduced by the Noise<sup>2</sup> vs. Signal method

### **Detector Discharge**

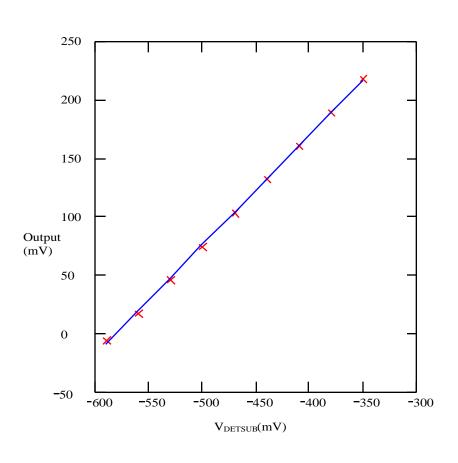


### **I-V Curves**



Typical I-V curve from one of the diodes (#14) from the 4th delivery, T=30K.

# Test Results for Delivery 4: Mux DC Gain

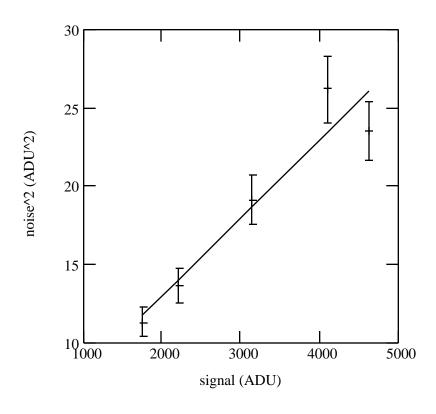


- turn off the reset switch
- discharge the detector
- measure output level vs.  $V_{DETSUB}$
- mux DC gain=0.95

### Capacitance

- $C = C_j + N \cdot \langle C_{mux} \rangle + C_{wire} + C_{bump}$ 
  - N number of mux pixels connected to the diode, of which
     1 is selected
  - <C<sub>mux</sub>> average capacitance of a single mux node (Rockwell estimates 0.04 pF for a selected node; capacitance of de-selected nodes larger)
  - C<sub>wire</sub> estimated to be 0.1 pF
  - $C_j$ -junction capacitance calculated from the depletion width, which is a function of the junction area,  $_s$  and  $N_D$ : ranges from 0.045 to 0.48 pF for the diodes tested
- C measured by the Noise<sup>2</sup> vs. Signal method yields higher values for  $< C_{mux} >$  than 0.04 pF, assuming the other values are either small or well-known

# Test Results for Delivery 4: Capacitance



Squared noise vs. signal for diode #14 from Delivery 4.

$$- C = C_j + N \cdot \langle C_{mux} \rangle + C_{wire} + C_{bump}$$

#### **Delivery #4**

 C for 5 diodes determined by the noise<sup>2</sup> vs. signal method, assuming Poisson Noise

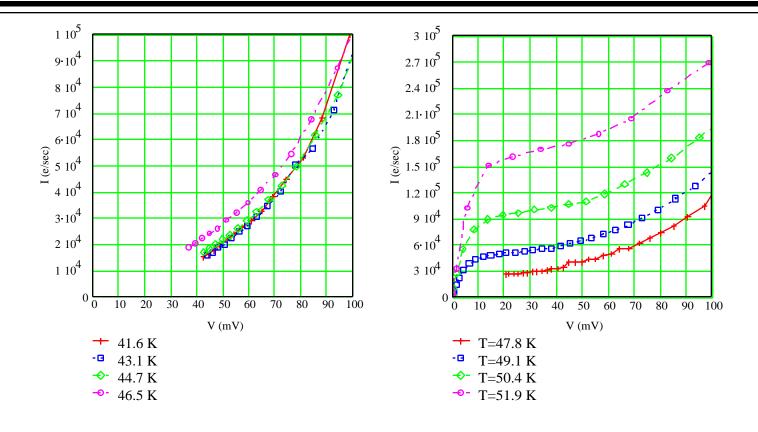
# Test Results for Delivery 4: Dark Current and R<sub>o</sub>A

- At 20 mV reverse bias, 9/14 diodes show dark current  $< 10^5$  e<sup>-</sup>/sec (20-40K),  $6/14 < 10^4$  e<sup>-</sup>/s at 30K
- 9/14 diodes show  $R_oA>10^7$  -cm<sup>2</sup>, many as high as  $2-8\,10^8$  -cm<sup>2</sup>

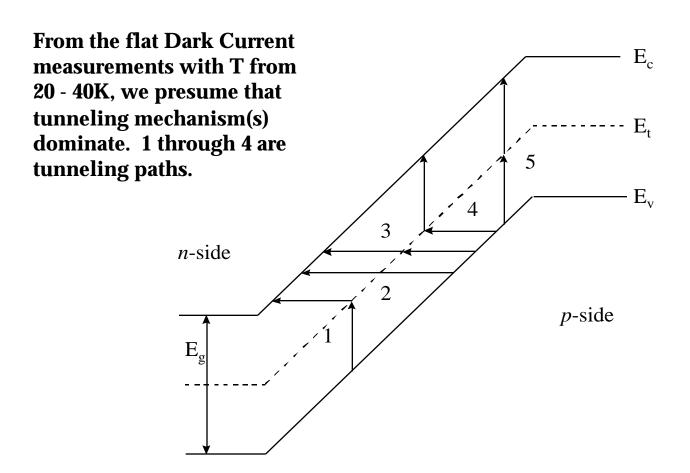
Above dark current and  $R_oA$  values are computed with C measured by the noise<sup>2</sup> vs. signal method

- Advanced detector architecture diodes show better performance
- Strong bias and weak temperature dependence: implies tunneling dominated dark currents.

### I-V Curves for Diode #15 as Function of T



# Band to Band (2), Trap-to-Band (1,3,4) and G-R (5) Dark Current Mechanisms



## Model: Trap-to-Band Tunneling

$$-I_{TB} = AN_{t}W \frac{^{2}qm^{*}EM^{2}}{h^{3}(E_{g} - E_{t})} exp - \frac{\sqrt{m^{*}/2}E_{g}^{3/2}F(a)}{2qE\hbar}$$

q - electron charge

A - junction area

N<sub>t</sub> - trap density

W - depletion width

 $m^*$  - effective mass of holes  $E_g$  - band gap energy

 $E_t$  - trap energy level (adjusted to fit data; = 0.037 eV)

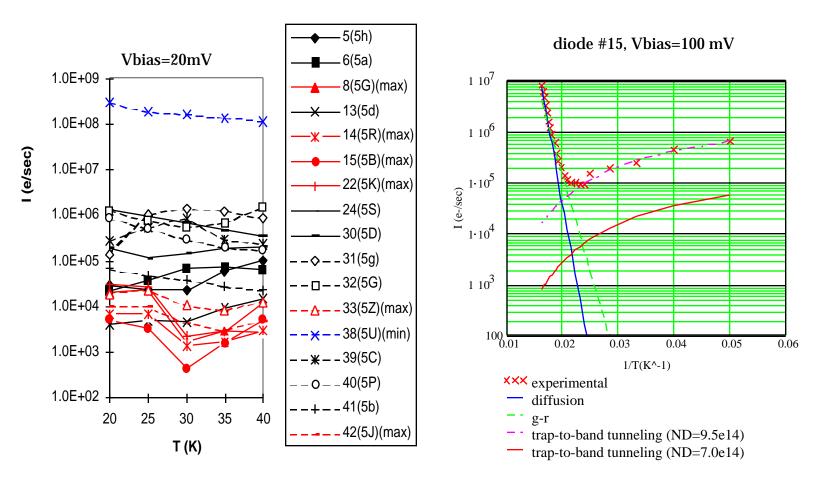
M - a matrix element associated with the trap potential

**F(a)** - defines the trap level distribution

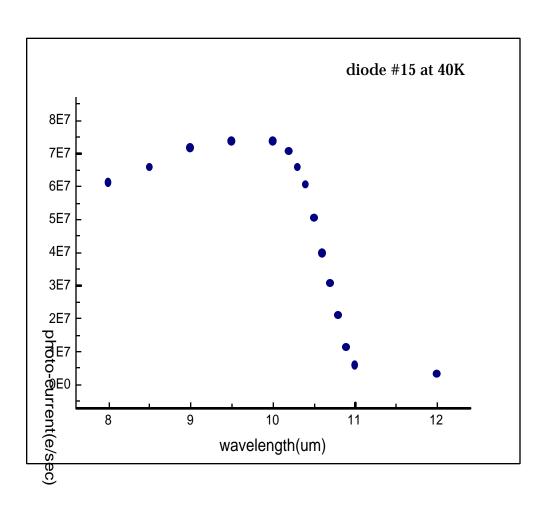
E - electric field at the junction and is proportional to  $N_D^{0.5}$ 

- Trap-to-band tunneling current increases rapidly with N<sub>D</sub>
- Model shows fit to data for 100 mV back bias temperature dependence results for #15, as well as expected improvement for lower N<sub>D</sub>

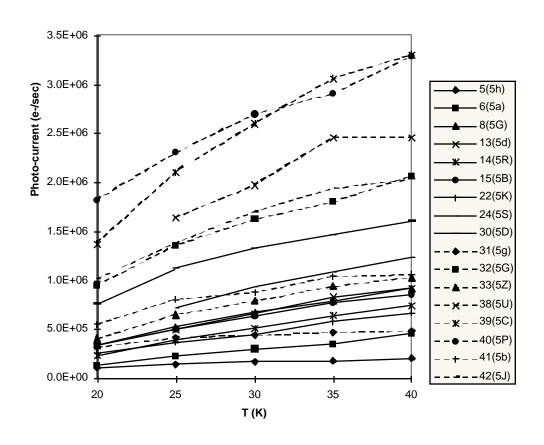
# Test Results for Delivery 4: Dark Current vs. Temperature



## Test Results for Delivery 4: Responsivity of Diode #15

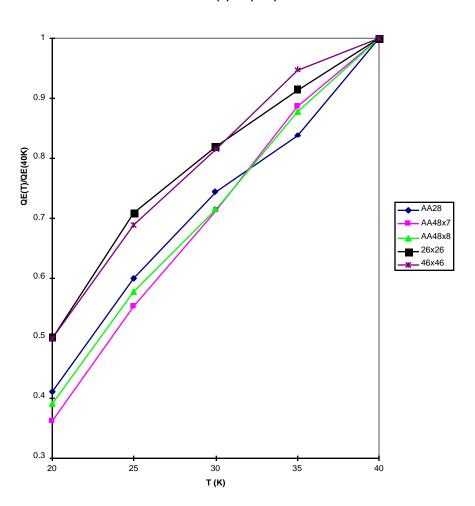


## Test Results for Delivery #4: Photocurrent Temperature Dependence



## Test Results for Delivery 4: Relative QE vs. Temperature

QE(T)/QE(40K)



- From 40 to 20 K, QE decreases by factor of 2 for most diodes
- QE of advanced architecture diodes has stronger temperature dependence

#### Discussion

• The diodes in delivery #4 show promise. Some diodes are close to satisfying space requirements [e.g. diode # 15 (5B), an advanced architecture diode, exhibits dark current < 440 e<sup>-</sup>/s at 30K and 20 mV back bias, and QE(30K) ~70% QE(40K)].

We conclude from this and other observations that advanced architecture diodes are superior:

- » have lower dark current
- » reasonable QE performance at lower temperatures
- Tunneling dominates the dark current, since dark current has strong bias dependence but weak temperature dependence. Will try material with lower  $N_{\rm D}$
- Deliveries 5 (and later) are focal plane arrays:
  - allow lower bias operation to minimize dark current
  - easier and more accurate capacitance measurement
  - back illumination, so that absolute QE measurements possible